

Noise Reduction Based on Modified Spectral Subtraction Method

Ekaterina Verteletskaya, Boris Simak

Abstract— In this paper, we propose a method for enhancing of speech corrupted by broadband noise. The method is based on the spectral subtraction technique. Performance of spectral subtraction, its limitations, artifacts introduced by it, and spectral subtraction modifications for eliminating these artifacts are discussed in the paper in details. To eliminate the musical noise, one of the artifacts introduced by conventional spectral subtraction, we propose to implement reduced varying scaling factor of spectral subtraction, with a following application of weighted function. Weighting function, used in the proposed algorithm, attenuates frequency spectrum components lying outside identified formants regions. The algorithm effects a substantial reduction of the musical noise without significantly distorting the speech. Listening tests were performed to determinate the subjective quality and intelligibility of speech enhanced by our method. Proposed noise reduction algorithm is compared to conventional spectral subtraction based on SNR improvement introduced by them. Spectrograms of speech, enhanced by the proposed algorithm and other modified spectral subtraction algorithms, which show the algorithms performance and degree of speech distortion, are also presented in the paper.

Index Terms— noise reduction, spectral subtraction, musical noise, noise estimation, VAD.

I. INTRODUCTION

Broadband noise presented in speech signal recorded under the real conditions can impair the quality of the signal, reduce intelligibility, and increase listener fatigue. Since in practice many kinds of noise is presented in recording speech, the problem of noise reduction is essential in the world of telecommunications and has gained much attention in recent years. Noise reduction or speech enhancement algorithms in general, attempt to improve the performance of communication systems when their input or output signals are corrupted by noise. The main objective of speech enhancement is to improve one or more perceptual aspects of speech, such as the speech quality or intelligibility. It is usually difficult to reduce noise without distorting speech and thus, the performance of speech enhancement systems is limited by the tradeoff between speech distortion and noise reduction. The complexity and ease of implementation of any proposed scheme is another important criterion especially since the majority of the speech enhancement and noise

reduction algorithms find applications in real-time portable systems like cellular phones, hearing aids, hands free kits etc.

Various classes of noise reduction algorithms have been developed mostly based on transform domain techniques, adaptive filtering, and model-based methods. Amongst the speech enhancement techniques, DFT-based transforms domain techniques have been widely spread in the form of spectral subtraction [1]. Even though the algorithm has very low computational complexity, it can reduce the background noise effectively. However, experimental results show that there is some residual noise in the processed signal, which affects the hearing effect. To reduce the influence of the background noise and increase the definition of the speech, the algorithm based on the modified spectral subtraction is introduced in this paper.

II. SPECTRAL SUBTRACTION METHOD

A. The principle of spectral subtraction

The spectral subtraction is based on the principle that the enhanced speech can be obtained by subtracting the estimated spectral components of the noise from the spectrum of the input noisy signal. Assuming that noise $d(n)$ is additive to the speech signal $x(n)$, the noisy speech $y(n)$ can be written as,

$$y(n) = x(n) + d(n), \text{ for } 0 \leq n \leq N-1 \quad (1)$$

Where n is the time index, N is a number of samples. The objective of speech enhancement is to find the enhanced speech $\hat{x}(n)$ from given $y(n)$, with the assumption that $d(n)$ is uncorrelated with $x(n)$. Input signal $y(n)$ is segmented into K segments of the same length. The time-domain signals can be transformed to the frequency-domain as,

$$Y_k(\omega) = X_k(\omega) + D_k(\omega), \text{ for } 0 \leq k \leq K-1 \quad (2)$$

Where k is the segment index, $Y_k(\omega)$, $X_k(\omega)$ and $D_k(\omega)$ denote the short-time DFT magnitudes taken of $y(n)$, $x(n)$, and $d(n)$, respectively, and raised to a power a ($a=1$ corresponds to magnitude spectral subtraction, $a=2$ corresponds to power spectrum subtraction). If an estimate of the noise spectrum \hat{D}_k can be obtained, then an approximation of speech \hat{X}_k can be obtained from Y_k

$$\hat{X}_k(\omega) = Y_k(\omega) - \hat{D}_k(\omega) \quad (3)$$

The noise spectrum cannot be calculated precisely, but can be estimated during period when no speech is present in the input signal. Most single channel spectral subtraction

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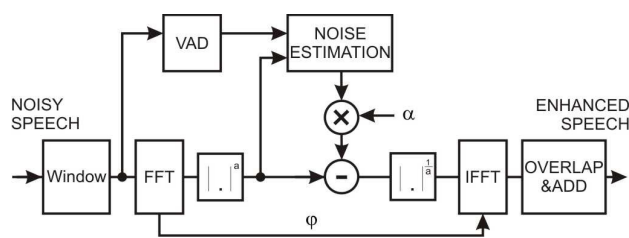


Figure 1. Spectral subtraction algorithm block-diagram

methods use a voice activity detector (VAD) to determine when there is silence in order to get an accurate noise estimate. The noise is assumed to be short-term stationary, so that noise from silent frames can be used to remove noise from speech frames.

Fig. 1 shows a block diagram of the spectral subtraction method. The extent of the subtraction can be varied by applying a scaling factor α [1]. The values of scaling factor α higher than 1 result in high SNR level of denoised signal, but too high values may cause distortion in perceived speech quality. Subtraction process with applying scaling factor α can be expressed as:

$$\hat{X}_k(\omega) = Y_k(\omega) - \alpha \cdot \hat{D}_k(\omega) \quad (4)$$

After subtraction, the spectral magnitude is not guaranteed to be positive. There are some possibilities to remove the negative components, for example half-wave rectification (setting the negative portions to zero), or full wave rectification (absolute value). Half-wave rectification is commonly used, but introduces musical tone artifacts in the processed signal. Full wave rectification avoids the creation of musical noise, but less effective at reducing noise. After subtraction, a root of the $\hat{X}_k(\omega)$ is extracted to provide corresponding Fourier amplitude components. An inverse Fourier transform, using phase components directly from Fourier transform unit, and overlap add is then done to reconstruct the speech estimate in the time-domain.

B. Noise estimation and voice activity detection (VAD)

A practical speech enhancement system consists of two major components, the estimation of noise power spectrum, and the estimation of speech. A critical component of any frequency domain enhancement algorithm is the estimation of the noise power spectrum. In single channel noise reduction/ speech enhancement systems, most algorithms require an estimation of average noise spectrum, and since a secondary channel is not available this estimation of the noise spectrum is usually performed during speech pauses. This requires a reliable speech/silence detector. The speech/silence detection scheme can be a determining factor for the performance of the whole system. The speech/silence detection is necessary to determine frames of the noisy speech that contain noise only. Speech pauses or noise only frames are used for the noise estimate updating, making the estimation more accurate.

The decision about voice activity presence is the sensitive part of the whole spectral subtraction algorithm as the noise power estimation can be significantly degraded by the errors in voice activity detection. VAD accuracy dramatically

affects the noise suppression level and amount of speech distortion that occurs. Many different techniques have been applied to the art of VAD. In the early VAD algorithms, short-time energy, zero-crossing rate, and linear prediction coefficients were among the common feature used in the detection process. Energy-based VADs [2] are frequently used because of their low computation cost. They work on principle, that the energy of the signal is compared with the threshold depending on the noise level. Speech is detected when the energy estimation lies over the threshold. Dynamical energy-based VAD described in [3] is used in proposed enhanced spectral subtraction method. In classical energy-based algorithms, detector cannot track the threshold value accurately, especially when speech signal is mostly voice-active and the noise level changes considerably before the next noise level re-calibration instant. The dynamical VAD was proposed to provide its classification more accurately in comparison with other energy-based techniques.

C. Limitations of spectral subtraction

1. *Impossibility to use for non-stationary noise.* Noise spectrum estimate is obtained from the speech nonnative regions of noisy speech. This assumption is valid for the case of stationary noise in which the noise spectrum does not vary much over time. Traditional VADs track the noise only frames of the noisy speech to update the noise estimate. But the update of noise estimate in those methods is limited to speech absent frames. This is not enough for the case of non-stationary noise in which the power spectrum of noise varies even during speech activity.
2. *Dependence on VAD accuracy.* Spectral subtraction performance is limited by the accuracy of noise estimation, which additionally is limited by the performance of speech/pause detectors [4]. Performance of whole spectral subtraction noise reduction algorithm as well as VAD performance degrades significantly at lower SNR.
3. *Musical noise.* Although spectral subtraction method provide an improvement in terms of noise attenuation, it often produce a new randomly fluctuating type of noise, referred to as musical noise due to their narrow band spectrum and presence of tone-like characteristics. This phenomenon can be explained by noise estimation errors leading to false peaks in the processed spectrum. When the enhanced signal is reconstructed in the time-domain, these peaks result in short sinusoids whose frequencies vary from frame to frame. Musical noise although very different from the original noise, can sometimes be very disturbing. A poorly designed spectral subtraction, which caused musical noise, can sometime results in the signal that has lower perceived quality and lower information content, than the original noisy signal. Most of the research at the present time is focused in ways to combat the problem of musical noise [5]. It is almost impossible to minimize musical noise without affecting the speech, and hence there is a tradeoff between the amount of noise reduction and speech distortion. Due to this fact several perceptual based

approaches were introduced, wherein instead of completely eliminating the musical noise (and introducing distortion), the noise is masked taking advantage of the masking properties of the auditory system [6].

III. MODIFICATIONS OF SPECTRAL SUBTRACTION

A. Spectral subtraction using scaling factor and spectral floor

The first spectral subtraction method proposed by Boll [7] consists of implementation of the following relationship:

$$|\hat{X}(\omega)|^2 = \begin{cases} |Y(\omega)|^2 - |\hat{D}(\omega)|^2, & \text{if } |Y(\omega)|^2 > |\hat{D}(\omega)|^2 \\ 0, & \text{otherwise} \end{cases} \quad (5)$$

As it was discussed above, though the noise is reduced by this method, there is still considerable broadband noise (musical noise) remaining in the processed speech. To eliminate this problem the method proposed in [1] introduces two additional parameters to basic spectral subtraction algorithm. There are scaling factor α , and spectral floor β . Since the residual noise spectrum consists of peaks and valleys with random occurrences, spectral subtraction using scaling factor and spectral floor tries to reduce the spectral excursions for improving speech quality. This proposed technique (see fig. 2) can be expressed as:

$$|\hat{X}(\omega)|^2 = \begin{cases} |Y(\omega)|^2 - \alpha \cdot |\hat{D}(\omega)|^2, & \\ \text{if } |Y(\omega)|^2 - \alpha \cdot |\hat{D}(\omega)|^2 > \beta \cdot |\hat{D}(\omega)|^2 & \\ \beta \cdot |\hat{D}(\omega)|^2, & \text{otherwise} \end{cases} \quad (6)$$

where $\alpha \geq 0$, and $\beta \ll 1$. The harshness of the subtraction can be varied by applying a scaling factor α . The values of scaling factor α higher than 1 result in high SNR level of denoised signal, but too high values may cause distortion in perceived speech quality. Therefore, the value of α has to be chosen carefully in order to prevent both the musical noise and signal distortion. The introduction of spectral floor prevents the spectral components of the enhanced speech spectrum to descend below the lower bound $\beta \cdot |D(\omega)|^2$, thereby “filling-in” the deep valleys surrounding narrow peaks (from the enhanced spectrum). Reducing the spectral excursions of noise peaks (as compared to when the negative components are set to zero) reduces the amount of musical noise.

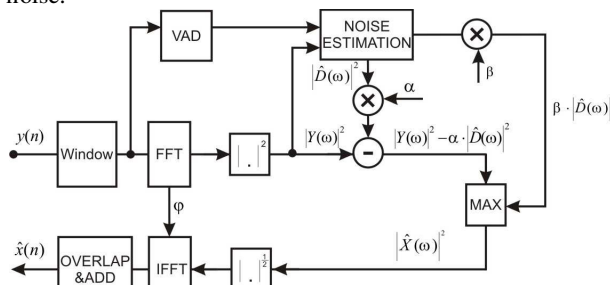


Figure 2. Block diagram of a modified spectral subtraction

The performance of this type of modified spectral subtraction algorithm is limited in the usage of stationary parameters, which are difficult to choose for all speech and noise situations. It is difficult to suppress noise without decreasing intelligibility and without speech distortion, especially for very low signal-to-noise ratios.

B. Wiener filtration

It is convenient to consider the spectral subtraction as a filter that can be expressed as the product of noisy speech signal spectrum and the frequency response of a spectral subtraction filter as:

$$|\hat{X}(\omega)|^2 = |Y(\omega)|^2 - |\hat{D}(\omega)|^2 = H(\omega) \cdot |Y(\omega)|^2 \quad (7)$$

$$H(\omega) = 1 - \frac{|\hat{D}(\omega)|^2}{|Y(\omega)|^2} = \frac{|Y(\omega)|^2 - |\hat{D}(\omega)|^2}{|Y(\omega)|^2} \quad (8)$$

The spectral subtraction filter is a zero phase filter, with its frequency response $H(\omega)$, which is in the range of $0 < H(\omega) < 1$. The filter acts as a SNR-dependent attenuator. The attenuation in each frequency increases with the decreasing SNR, and vice-versa.

A transfer function of the Wiener filter [8], $H_{wiener}(\omega)$, is expressed in terms of the power spectrum of clean speech $P_x(\omega)$ and the power spectrum of noise $P_d(\omega)$ as in (9). But power spectrum of clean speech is not known, the power spectrum of the noisy speech $P_y(\omega)$ signal is used instead as:

$$H_{wiener}(\omega) = \frac{P_x(\omega)}{P_x(\omega) + P_d(\omega)} = \frac{P_y(\omega) - P_d(\omega)}{P_y(\omega)} \quad (9)$$

Wiener filter cannot be applied directly to estimate the clean speech since speech cannot be assumed to be stationary. Therefore, an adaptive Wiener filter implementation can be used to approximate the above filter (10) as:

$$H_{wiener}(\omega) = \frac{E[|Y(\omega)|^2] - E[|D(\omega)|^2]}{E[|Y(\omega)|^2]} \quad (10)$$

$$|\hat{X}(\omega)|^2 = H_{wiener}(\omega) \cdot |Y(\omega)|^2 \quad (11)$$

Comparing $H(\omega)$ and $H_{wiener}(\omega)$ from (8) and (10), it can be seen that the Wiener filter is based on the *ensemble average spectra* of the signal and noise, whereas the spectral subtraction filter uses the instantaneous spectra for noise signal and the running average (*time-averaged spectra*) of the noise. In Wiener filter theory the averaging operations are taken across the ensemble of different realization of the signal and noise processes. In spectral subtraction we have access only to single realization of the process.

Using of power spectrum of noisy speech, instead of that of clean speech for calculating the transfer function degrades Wiener filter accuracy. To solve this problem, an iterative algorithm is used [8]. In the algorithm the output signal of the Wiener filter is utilized to design a more accurate Wiener filter. Thus by iterating this process, we can design a high accurate Wiener filter. The input signal of the iterative Wiener filter is not renewed at each iteration, only the filter is renewed.

C. Iterative spectral subtraction

To consider the musical noise problem common to conventional spectral subtraction method, an iterative

spectral subtraction method was proposed in [9]. The iterative method is motivated by iterative Wiener filtering, where filtering output signal is used to design a higher performance Wiener filter. In iterative spectral subtraction the filtering output signal is used not only for designing the filter but also as the input signal of the next iteration process. Specifically for spectral subtraction, after the first spectral subtraction process, the type of additive noise is changed to that of musical noise. Then the noise signal is estimated from unvoiced segment parts. And, a new spectral subtraction filter is designed by using the new estimated noise (musical noise) and the new noisy speech (including the musical noise), which is the output signal by the first spectral subtraction. By the designed filter, we can obtain an enhanced output signal from the input signal. At every iteration musical noise is estimated in different frames, because the musical noise is not stationary in short time frames analysis. When we do such noise estimation, the spectral subtraction filter is always designed so as to reduce the musical noise remained in the previous spectral subtraction process. Therefore, the musical noise can be reduced significantly by performing the iterative spectral subtraction. The best result is obtained with the number of iterations equals to 2 [9].

D. Spectral subtraction based on perceptual properties

The choice of the subtraction parameters α , β and a is the main challenge in subtractive type speech enhancement algorithms. To track changes in background noise it is necessary to subtraction parameters to be adaptive. Good results are obtained, when the adaptation of subtractive parameters in time and frequency domain based on masking properties. Masking consists in the fact, that the human auditory system does not distinguish two signals when the signals are close to each other (in the time or frequency domain). In [10] the noise masking threshold $T(\omega)$ is used for adjusting spectral subtraction parameters α and β on a per frame and per frequency basis. The noise masking threshold is obtained through modeling the frequency selectivity of the human ear and its masking property. The different calculation steps of noise masking threshold are summarized in [10].

Spectral subtraction algorithms based on perceptual properties are supposed to have the best results among all modified spectral subtraction algorithms. In case of this type of SS small amount of residual noise is leaved, but this noise has a perceptually white quality and distortion remains acceptable. One of the disadvantages of SS based on perception properties is their computational complexity. In proposed method we tried to combine advantages of spectral subtraction based on perceptual properties with the computational simplicity.

IV. PROPOSED METHOD

A. Weighting function

As already discussed above, the spectral subtraction technique employed in the apparatus of Fig.1 has the disadvantage that output, though less noisy than the input signal, contains musical noise. The majority of information in a segment of noise-free speech is contained within one or

more high energy frequency bands, known as formants. Within the formant regions themselves, the musical noise is largely masked out by the speech itself. Proposed spectral subtraction algorithm [11] aims to reduce the audible musical noise by attenuating the signal in the regions of the frequency spectrum lying between the formant regions. Attenuation of the regions between the formants has little effect on the perceived quality of the speech itself, so that this approach is able to effect a substantial reduction in the musical noise without significantly distorting the speech. This attenuation is performed by weight function $W(\omega)$ derived from the spectral envelope $L(\omega)$, which is obtained by means of linear prediction analysis.

B. Implementation

The overall block diagram of proposed enhanced spectral subtraction method is shown on Fig.3. The noisy speech is segmented into overlapping frames. Then Hamming window is applied on each segment and a set of Fourier coefficients using short-time fast Fourier transform (FFT) is generated. To get the spectrum of noisy speech, Y_k , Fourier coefficients are raised to a power a ($a=1$ or 2). Noise spectrum is estimated during periods when no speech is present in the input signal. This condition is recognized by voice activity detector (VAD) to produce a control signal which permits the updating of store with spectrum Y_k when speech is absent from the current segment. This spectrum is smoothed by making each frequency samples of Y_k the average of adjacent frequency samples, given \hat{Y}_k . This smoothed spectrum then will be used to update a spectral estimate of noise, which consists of a proportion of the previous noise and a portion of the smoothed short-term spectrum of current segment. Thus the noise spectrum gradually adapts to changes in the actual spectrum noise. It can be defined as

$$\hat{D}_k = \lambda \cdot \hat{D}_{k-1} + (1 - \lambda) \cdot \hat{Y}_k \quad (12)$$

Where \hat{D}_k is the updated noise spectral estimate, k is a frame index, \hat{D}_{k-1} is the old noise spectral estimate, \hat{Y}_k is the

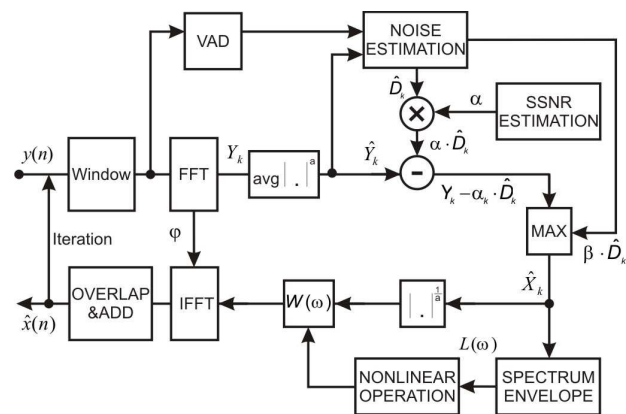


Figure 3. Block-diagram of proposed algorithm

smoothed noise spectrum from the present frame, and λ is a decay factor. Present noise spectral estimate is subtracted from the noisy speech power spectrum:

$$\hat{X}_k = Y_k - \alpha_k \cdot \hat{D}_k \quad (13)$$

The scaling factor α_k defines a subtraction dimension. When the spectral subtraction is followed by the weighting function $W(\omega)$ a lower value of the scaling factor can be used, therefore speech will be less distorted. The value of α_k in proposed method is the function of segmental SNR and varies from frame to frame within the same signal. Using the SSNR α_k can be determined for each frame as:

$$\alpha_k = \begin{cases} 3, & SSNR_k < 0dB \\ 3 - SSNR_k / 10, & 0dB \leq SSNR_k \leq 20dB \\ 1, & SSNR_k > 20dB \end{cases} \quad (14)$$

Subtraction defined in (13) may result in negative terms. Since a frequency component cannot have a negative power, in proposed method a nonzero minimum noise power level $Z_k = \beta \cdot \hat{D}_k$ is defined, where β determinates the minimum power level or "spectral floor". Thereby output of spectral subtraction \hat{X}_k is defined as the maximum of $Y_k - \alpha_k \cdot \hat{D}_k$ and $\beta \cdot \hat{D}_k$. A non zero value of β ($0 < \beta < 1$) reduces the effect of musical noise by retaining a small amount of the original noise signal.

Musical noise in proposed method is also reduced by attenuating the signal in the regions of the frequency spectrum lying between the formant regions. This attenuation is performed by special unit, which multiplies the Fourier coefficients by respective terms of a weighting function $W(\omega)$. The weighting function is obtained from spectral envelope $L(\omega)$ which is obtained by means of a LPC analysis. The attenuation operation is such that any coefficient of the spectrally subtracted speech \hat{X}_k is attenuated only if the corresponding frequency term of the spectral envelope is below a threshold value τ . Thus the response $W(\omega)$ is a nonlinear function of $L(\omega)$ and is obtained by nonlinear processing unit according to the rule:

$$\begin{aligned} & \text{if } L(\omega) \geq \tau, W(\omega) = 1 \\ & \text{else } W(\omega) = \left[\frac{L(\omega)}{\tau} \right]^\gamma \end{aligned} \quad (15)$$

The threshold value τ is a constant for all frequencies and for all speech segments. In a strongly voiced segment of speech, only small portions of the spectrum will be attenuated, whereas in quiet segments most of the spectrum may be attenuated. A value of the threshold about 10% of peak amplitude of the speech is found to work well. A lower value of τ will produce a more harsh filtering operation. Thus the value could be increased for higher SNR, and lowered for lower signal to noise ratios. The power term $\gamma=2$ were used. Larger value of γ will make the attenuation harsher. The

value of γ may be used to vary the harshness of the attenuation.

After subtraction and nonlinear weighting, the a root of the output terms is taken to provide corresponding Fourier amplitude components, and the time-domain signal segments reconstructed by an inverse Fourier transform unit from these along with phase components ϕ directly from the FFT unit. The windowed speech segments are overlapped to provide the reconstructed output signal at an output.

We increase algorithm performance by iteration of whole process [9]. After the first spectral subtraction process, the type of additive noise is changed to that of musical noise. We estimate the noise signal from unvoiced segment parts using the VAD. Noise estimate accuracy increases due to VAD accuracy increasing after first spectral subtraction, which results in SNR improvement. We design a new spectral subtraction by using the new estimated noise (musical noise) and the new noisy speech (including the musical noise), which is the output signal by the first spectral subtraction. Therefore, musical noise also can be reduced by performing the iteration of spectral subtraction as shown on Fig. 3.

V. EXPERIMENTS AND DISCUSSION

A. Speech data and analysis parameters

Proposed method has been tested on real speech data by computer simulation in MATLAB environment. Real speech signals from SpeechDat database were used for experiments. The sampling frequency of the signals is 8kHz. Utterances were pronounced by female and male speakers in Czech language. Three types of noise were used to generate noisy speech signals with the different SNR level ($SNR = [0dB \ 5dB \ 10dB \ 15dB]$). The noises used for experiments are following: AWGN noise generated by computer, office noise, and engine noise recorded inside the car taken from CAR2ECS database [12]. Parameters for computing are as follows. The frame length is 20ms (160 samples). Each frame is 50 % overlapped. Frames were windowed with Hamming window and 256 points FFT is applied to each frame. We have experimented with two values of power a ($a=1, a=2$), and with or without iteration of proposed algorithm. The value $a=2$ and one iteration of whole algorithm were found to result best algorithm performance.

The proposed noise reduction method was compared to conventional spectral subtraction based on SNR improvement introduced by them. In addition, subjective speech quality and intelligibility evaluation tests have been held to evaluate the quality of noisy speech enhanced by conventional spectral subtraction and proposed noise reduction method.

The difference in performance of all above discussed noise reduction techniques based on spectral subtraction and speech distortion introduced by them can be seen on spectrograms of speech utterance enhanced by these algorithms.

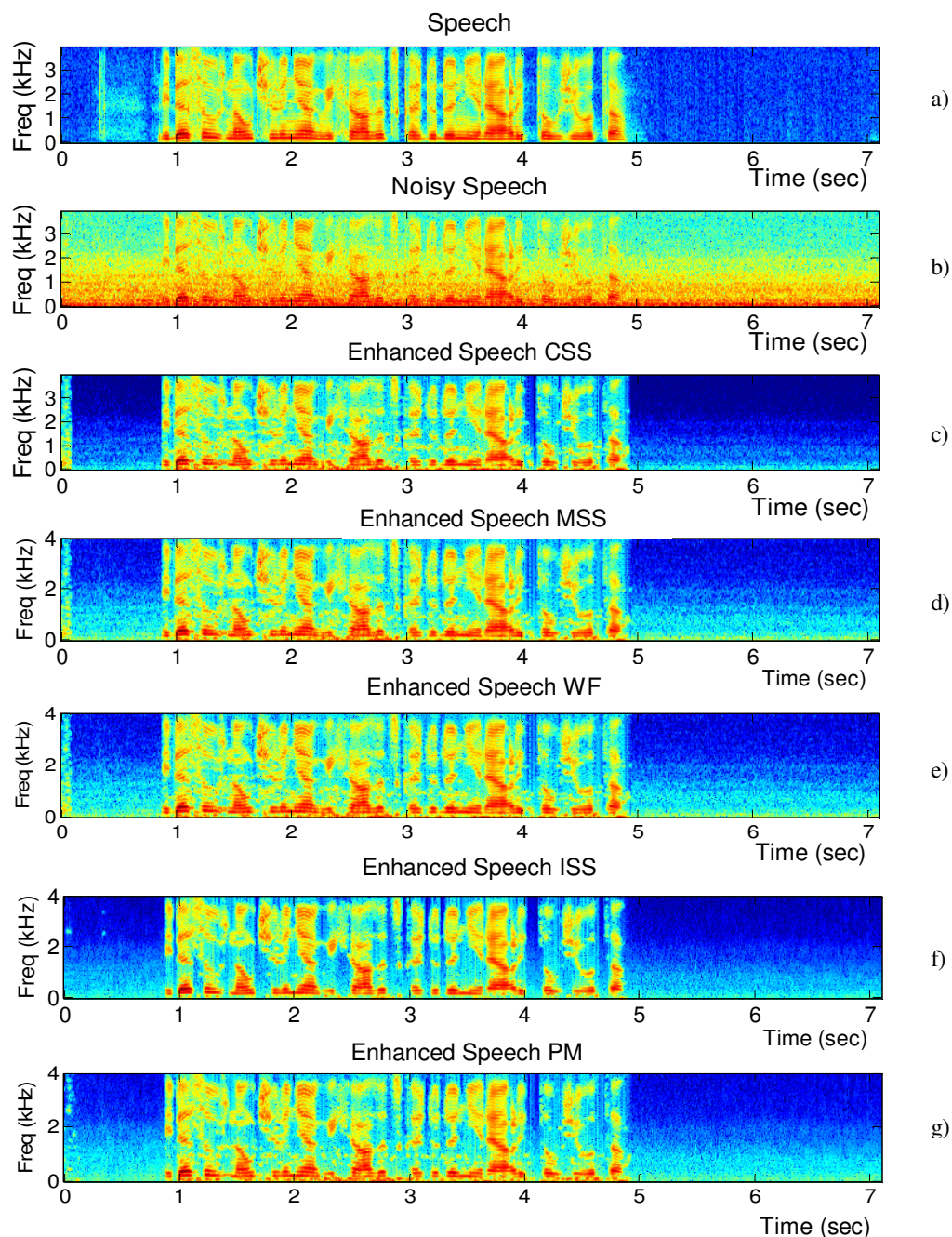


Figure 4. Speech spectrograms. (a) Clean speech, (b) Noisy speech in the case of additive car noise (SNR = 0 dB), (c) – (g) Speech enhanced by noise reduction algorithms. (c) CSS -conventional spectral subtraction, (d) MSS - modified spectral subtraction using scaling factor and spectral floor, (e) WF – Wiener filtration, (f) ISS – iterative spectral subtraction, (g)PM – proposed method

B Numerical results

The results of the proposed noise reduction algorithm, as well as comparison to conventional spectral subtraction algorithm are shown on Fig. 5. Evaluation of the algorithm performance for different kinds of noise and different levels of SNR is indicated by an improvement of segmental SNR (SSNR). The SSNR is a mean value of SNR in each frame. The improvement of SSNR is obtained by the SSNR of the output signal minus the segmental SNR of the input signal.

Fig. 5b presents outcomes of subjective speech quality and intelligibility test. In the experiments, we adopt informal speech test. Five noisy speech sentences were processed by two methods (conventional spectrum subtraction (CSS) and

the method proposed in this paper (PM) respectively). Five listeners were invited to test and evaluate quality and intelligibility of enhanced speech sentences for three times. As it shown on Fig.5 the best results in SNR improvement of noisy speech were obtained by proposed algorithm on AWGN noise. The difference in SNR improvement between conventional SS and proposed method increase considerably with the decreasing input signal SNR conditions. For car and office noise in 0dB SNR level the difference in SNR improvement between two concerned algorithms is 1.5-2dB. However the difference in speech quality and intelligibility enhancement between conventional SS and proposed algorithm is considerable.

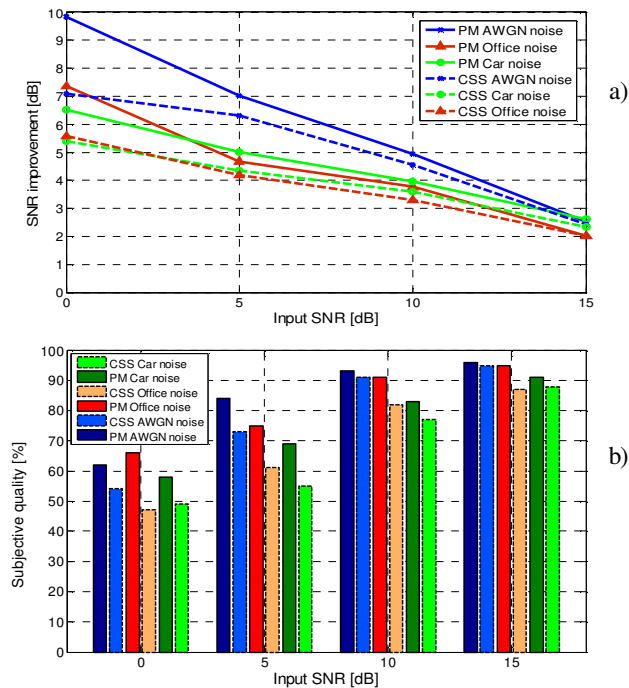


Figure 4. SNR improvement (a) and speech quality assesment (a) of conventional SS (CSS) and proposed method (PM) for various noise types and input signal SNR levels

The main drawback of the SNR is the fact that it has a very poor correlation with subjective results. SNR of enhanced speech is not sufficient objective indicator of speech quality. Structure of residual noise and speech distortion can be seen on spectrograms of enhanced speech. Fig.4 represents spectrograms of speech enhanced by above described algorithms (conventional spectral subtraction (CSS), modified spectral subtraction (MSS) with scaling factor and spectral floor, Wiener filtration (WF), Iterative spectral subtraction (ISS) and proposed algorithm (PM) based on perceptual weighting). As it shown on Fig.4 conventional SS as well as modified SS contain audible residual noise, which can be annoying for listener. Wiener filtering results in a smaller amount of residual noise, but this noise has musical structure and decreases speech intelligibility. In case of iterative SS background noise is well attenuated, but some of speech regions, especially fricative consonants, are also attenuated. This type of SS can result in speech distortion. The best results were obtained with proposed algorithm with perceptual weighting. In this case small amount of residual noise is leaved, but this noise has a perceptually white quality and distortion remains acceptable.

In low SNR conditions conventional SS algorithm results in annoying musical noise and some speech distortions. In spite of high factor of SNR improvement, enhanced signal has a lower perceived quality and lower information content, than the original noisy signal. Proposed algorithm even in low input signal SNR levels result in high results of SNR improvement as well as in substantial speech quality enhancement with minimal distortions (see Fig.4).

VI. CONCLUSION

In this paper speech enhancing method based on improved spectral subtraction algorithm was introduced. For effective noise reduction with minimal distortion proposed algorithm takes in account perceptual aspects of human ear. It can be seen from the experimental results that proposed method effectively reduces background noise in comparison with commonly used spectral subtraction type algorithms. Proposed method results in greater improvement of SNR and considerably improvement of perceptual speech quality in compasion to conventional spectral subtraction method.

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